

## z-Transform Review

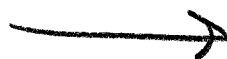
(6-1)

- you should already be familiar with the z-transform, its relationship to the DTFT, and its relationship to the Laplace transform from your undergraduate signals and systems course.
- The z-transform is treated in chapter 6 of the text.
- Here we'll just do a brief review.

## Eigenvalue Interpretation

FACT: for any fixed complex number  $z$ , the signal  $z^n$  is an eigenfunction of any discrete time LTI system.

Proof: Let  $H$  be an LTI system with impulse response  $h[n]$ . Let  $z \in \mathbb{C}$  be a constant and let the system input be  $z^n$ .



Then the system output is given by.

(6-2)

$$y[n] = H\{z^n\} = z^n * h[n]$$

$$= \sum_{k \in \mathbb{Z}} h[k] z^{n-k} = \sum_{k \in \mathbb{Z}} h[k] z^n z^{-k}$$

$$= z^n \sum_{k \in \mathbb{Z}} h[k] z^{-k} = H(z) z^n.$$

The input

A number that depends on the value of the constant  $z$  and on the system

QED.

- So we see that, for the set of eigenfunctions  $\{z^n\}_{z \in \mathbb{C}}$ , the eigenvalues are given by the  $z$ -transform of the impulse response.

-  $H(z)$  is called the "transfer function" of the system.

## The "Fixer-Upper" Interpretation

6-3

- Another way to think of the z-transform that is probably more intuitive for most people.
- We already talked about this in the notes for chapter 1.
- Suppose we want to consider signals  $x[n]$  that may have divergent "bad behaviour" on the right side (as  $n \rightarrow \infty$ ) or on the left side (as  $n \rightarrow -\infty$ ).

EX:  $x[n] = 3^n$  or  $x[n] = 3^n u[n]$  ;  
bad on the right.

$x[n] = (\frac{1}{2})^n$  or  $x[n] = (\frac{1}{2})^n u[-n]$  ;  
bad on the left.

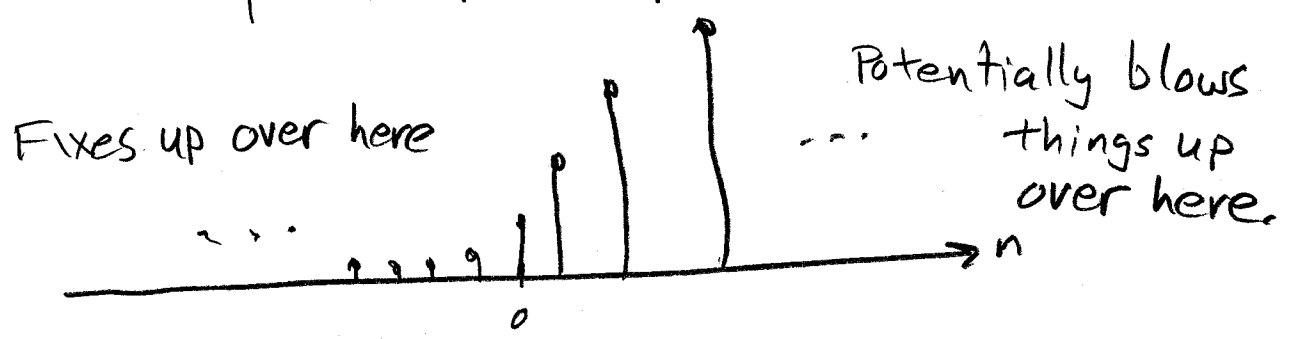
- The DTFT fails to converge for signals of this type.
- If we want to be able to use transform analysis to analyze how such signals interact with LTI systems, we will need a transform that is more powerful than the DTFT.

- The strategy is to multiply the "bad" signal  $x[n]$  times a "fixer-upper" signal  $r^{-n}$ , where  $r \in \mathbb{R}$  and  $r > 0$ . 6-4

- The fixer-upper corrects the bad behaviour in  $x[n]$  so that we can take the DTFT of the "fixed up" signal  $x[n]r^{-n}$ .

→ if  $0 < r < 1$ , then the fixer-upper  $r^{-n}$  fixes up on the left.

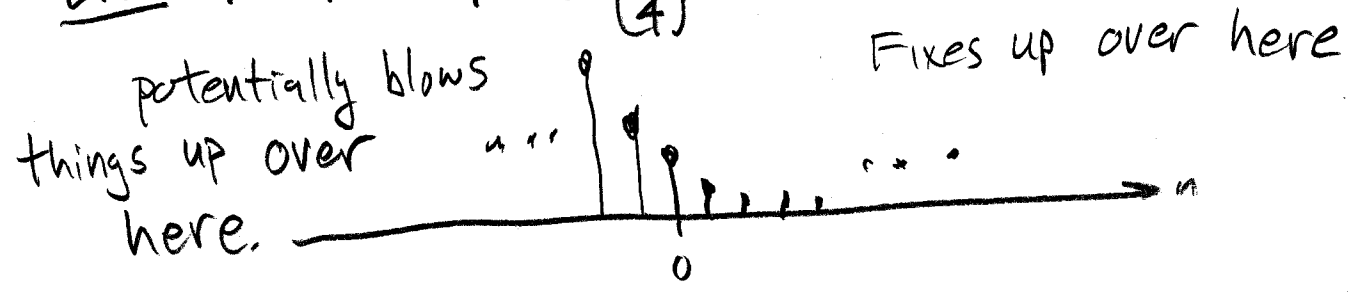
EX:  $r = \frac{1}{4} \rightarrow r^{-n} = 4^n$  :



→ If  $r = 1$ , then  $r^{-n} = 1$  and the fixer-upper does nothing: the fixed up signal is  $x[n]r^{-n} = x[n]$ .

→ If  $r > 1$ , then  $r^{-n}$  fixes up on the right.

EX:  $r = 4 \rightarrow r^{-n} = (\frac{1}{4})^n$



- The  $z$ -transform of  $x[n]$  is the DTFT of the fixed-up guy  $x[n]r^{-n}$ :

(65)

$$X(z) = \text{DTFT} \{ x[n]r^{-n} \}$$

$$= \sum_{n \in \mathbb{Z}} x[n]r^{-n} e^{-j\omega n}$$

$$= \sum_{n \in \mathbb{Z}} x[n] \underbrace{(re^{j\omega})^{-n}}$$

a complex number in polar form,  
call it  $z$ , with magnitude  
 $r = |z|$  and angle  $\omega = \arg z$ .

$$= \sum_{n \in \mathbb{Z}} x[n]z^{-n}$$

- You can think of the graph of  $X(z)$  as a sheet or surface that exists over the complex  $z$ -plane.

- Actually, since  $X(z)$  is complex valued in general, it takes "two graphs" to show  $X(z)$ :

- one for the real part and one for the imaginary part,
- or one for the magnitude and one for the phase [e.g., the angle].

- There is a nice example of one of these type graphs in Fig. 6.2 on page 283 of the text. 6-6

- The set of complex numbers  $z$  for which  $X(z)$  converges is called the "Region of Convergence" or "ROC".

- Complex numbers  $z$  for which  $X(z)$  blows up (diverges) are not in the ROC.

- These  $z$  are "singularities" of  $X(z)$ .

- Singularities that occur at isolated, single points in the  $z$ -plane are called "poles".

- For each real  $r \geq 0$ , the DTFT of the fixed up guy  $x[n]r^{-n}$  either converges or it does not.

- If it converges, then the circle of radius  $r$  in the  $z$ -plane is included in the ROC of  $X(z)$ ,

→ And the graph of the DTFT of  $x[n]r^{-n}$  is part of the graph of  $X(z)$ .

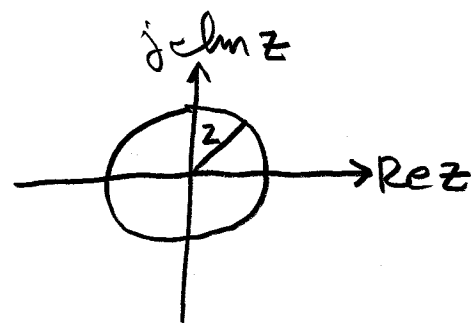


EX: Suppose that

(6-7)

$$X(z) = \sum_{n \in \mathbb{Z}} x[n] z^{-n} = \sum_{n \in \mathbb{Z}} x[n] r^{-n} e^{-j\omega n}$$

converges on the circle  $|z|=2$ .

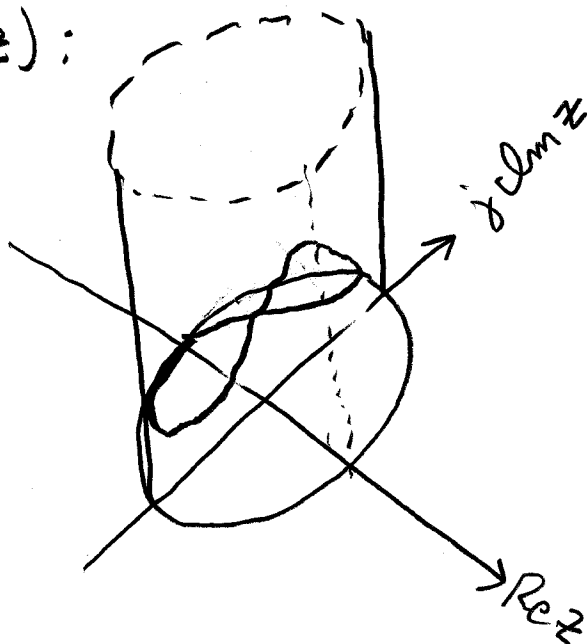


- Then the

graph of the DTFT of the fixed up guy  $x[n] z^{-n} = (\frac{1}{2})^n x[n]$

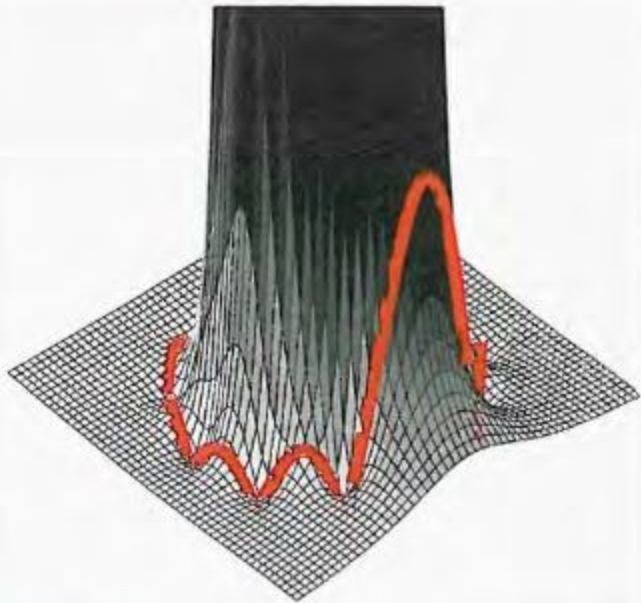
is wrapped around this circle as part of the graph of  $X(z)$ :

$$\text{DTFT} \left\{ \left(\frac{1}{2}\right)^n x[n] \right\}$$

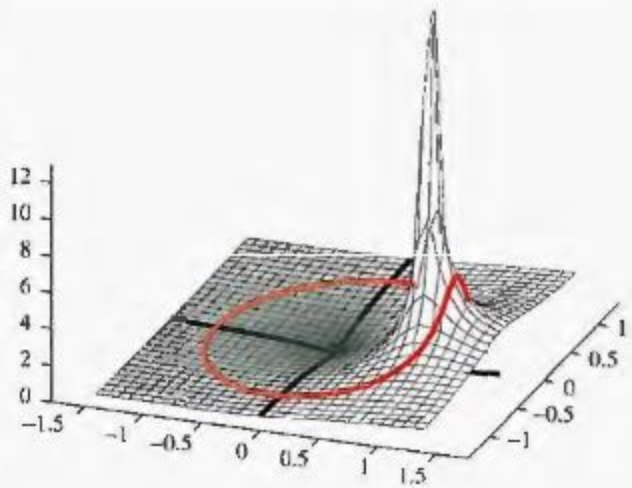


- So you can think of the ROC of  $X(z)$  as a collection of circles of radius  $r$  where the DTFT of  $x[n] r^{-n}$  converges.









- If  $x[n]$  is zero for all the positive  $n$ 's, then (6-8)

$$X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n} = \dots + x[-2]z^2 + x[-1]z^1 + x[0]z^0 + 0 \frac{1}{z} + 0 \frac{1}{z^2} + \dots$$

and the ROC will generally include the circle of radius 0, e.g. the point  $z=0$ .

- If  $x[n]$  is zero for all the negative  $n$ 's, then the ROC will generally include the circle of radius  $\infty$ , e.g. the set  $|z| = \infty$ .

- You can think of the  $z$ -transform  $X(z)$  as a collection of DTFT's of fixed-up guys  $x[n]r^{-n}$  wrapped around concentric circles of radius  $r$  in the  $z$ -plane.

- For  $r=1$ , the fixed up guy is  $x[n]1^n = x[n]$ . So, above the unit circle in the  $z$ -plane  $X(z)$  is exactly identical to  $X(e^{j\omega})$ .

- If  $x[n]$  has a DTFT, then the unit circle is included in the ROC of  $X(z)$ .

- If  $x[n]$  does not have a DTFT, then the unit circle is not included in the ROC of  $X(z)$ .

NOTE: when  $r=1$ , then  $z = re^{j\omega} = e^{j\omega}$ ,  
so  $X(z) = X(e^{j\omega})$ .

- In order to specify a z-transform  $X(z)$ , it's necessary to specify both the function  $X(z)$  and the ROC which tells which  $z$ 's  $X(z)$  converges for.

EX:

$$x_1[n] = \left(\frac{1}{2}\right)^n u[n].$$

$$\begin{aligned} X_1(z) &= \sum_{n \in \mathbb{Z}} \left(\frac{1}{2}\right)^n u[n] z^{-n} = \sum_{n=0}^{\infty} \left(\frac{1}{2}\right)^n z^{-n} \\ &= \sum_{n=0}^{\infty} \left(\frac{1}{2} z^{-1}\right)^n = \frac{1}{1 - \frac{1}{2} z^{-1}}, \quad |z| > \frac{1}{2} \end{aligned}$$

NOTE: for  $|z| \leq \frac{1}{2}$ , the z-transform

$X_1(z)$  is NOT equal to  $\frac{1}{1 - \frac{1}{2} z^{-1}}$ . For

$|z| \leq \frac{1}{2}$ ,  $X_1(z)$  diverges to  $\infty$ .



EX ... Now let

(6-10)

$$x_2[n] = -\left(\frac{1}{2}\right)^n u[-n-1].$$

Then  $X_2(z) = \sum_{n \in \mathbb{Z}} -\left(\frac{1}{2}\right)^n u[-n-1] z^{-n}$

$$= -\sum_{n=-\infty}^{-1} \left(\frac{1}{2}\right)^n z^{-n} = -\sum_{n=1}^{\infty} 2^n z^n$$

$$= 0 - \sum_{n=1}^{\infty} 2^n z^n = 1 - 1 - \sum_{n=1}^{\infty} 2^n z^n$$

$$= 1 - 2^0 z^0 - \sum_{n=1}^{\infty} 2^n z^n$$

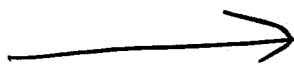
$$= 1 - \sum_{n=0}^{\infty} 2^n z^n = 1 - \frac{1}{1-2z}, \quad |z| < \frac{1}{2}$$

$$= \frac{1-2z}{1-2z} - \frac{1}{1-2z}, \quad |z| < \frac{1}{2}$$

$$= \frac{1-2z-1}{1-2z}, \quad |z| < \frac{1}{2}$$

$$= \frac{2z}{2z-1} \cdot \frac{1/2}{1/2} = \frac{z}{z-1/2} \cdot \frac{z^{-1}}{z^{-1}}$$

$$= \frac{1}{1-\frac{1}{2}z^{-1}}, \quad |z| < \frac{1}{2}.$$



EX ... So we have

(6-11)

$$X_1(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}, \quad |z| > \frac{1}{2}$$

and

$$X_2(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}, \quad |z| < \frac{1}{2}.$$

- Are  $X_1(z)$  and  $X_2(z)$  equal?

→ NO!!!!

→ There is no number  $z \in \mathbb{C}$  for  
which  $X_1(z) = X_2(z)$  !!!

→ For any  $z$  that  $X_1(z)$  converges to  
the number  $\frac{1}{1 - \frac{1}{2}z^{-1}}$ ,  $X_2(z)$  diverges!

→ For any  $z$  that  $X_2(z)$  converges to the  
number  $\frac{1}{1 - \frac{1}{2}z^{-1}}$ ,  $X_1(z)$  diverges!

⇒ The ROC is important.

(6-12)

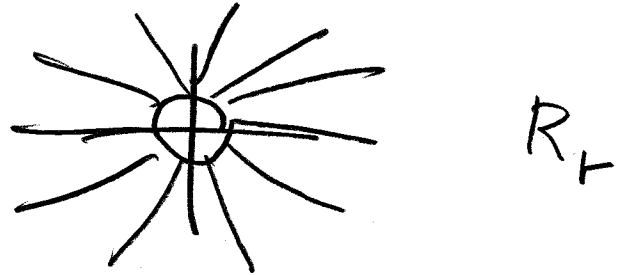
- If  $x[n]$  is right sided, then any bad behavior could only be on the right side.
- If there is any  $r_0$  that can fix things up for this  $x[n]$ , say  $r_0 = 2$ , then  $(\frac{1}{2})^n x[n]$  has a convergent DTFT.
- For any  $r > r_0$ , like  $r = 3$  in this case, the guy  $x[n]$  is fixed up even more. The DTFT of  $(\frac{1}{3})^n x[n]$  certainly converges, because  $(\frac{1}{3})^n$  fixes up bad behavior on the right even more than  $(\frac{1}{2})^n$ .

⇒ A right-sided  $x[n]$  has an exterior ROC. It is generally the exterior of a circle that passes through the largest pole of  $X(z)$ .

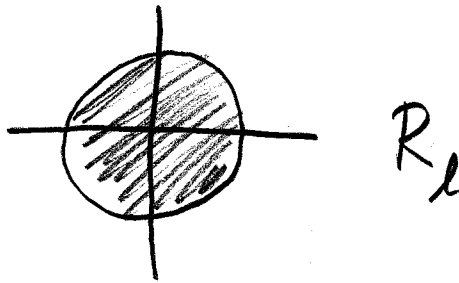
⇒ A left-sided  $x[n]$  has an interior ROC. It is generally the interior of a circle that passes through the smallest pole of  $X(z)$ . →

- Any two sided  $x[n]$  can be broken into 6-13  
the sum of a right sided part plus a left  
sided part,

- The  $z$ -transform of the right sided part has  
a ROC that is the exterior of a circle  
in the  $z$ -plane,

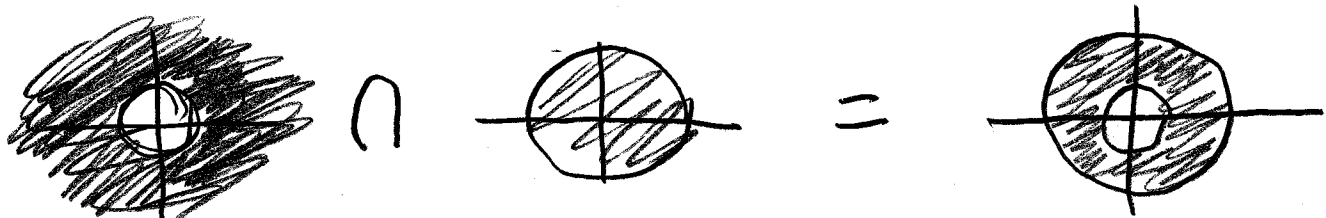


- The  $z$ -transform of the left sided part has a  
ROC that is the interior of a circle in the  
 $z$ -plane.



- The ROC of  $X(z)$ , the  $z$ -transform of the whole  
signal  $x[n]$ , is exactly the set of  $z$ 's for which  
both the  $z$ -transform of the right sided part  
and the  $z$ -transform of the left sided part  
converge.

- The ROC of  $X(z)$  is an annulus.



⇒ A two-sided  $x[n]$  has an annular ROC. (6-14)

⇒ An  $x[n]$  with finite support (finite length) has a ROC that is generally the entire  $z$ -plane, except possibly the points  $z=0$  and  $z=\infty$ .

→ These two points must be checked individually for each  $X(z)$ .

$$\rightarrow X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

$$= \dots + x[-2] z^2 + x[-1] z^1 + x[0] z^0 + x[1] \frac{1}{z} + x[2] \frac{1}{z^2} + \dots$$

→ If  $x[n] = 0 \quad \forall n > 0$ , then the point  $z=0$  is included in the ROC of  $X(z)$ .

→ If  $x[n] = 0 \quad \forall n < 0$ , then the point  $z=\infty$  is included in the ROC of  $X(z)$ .

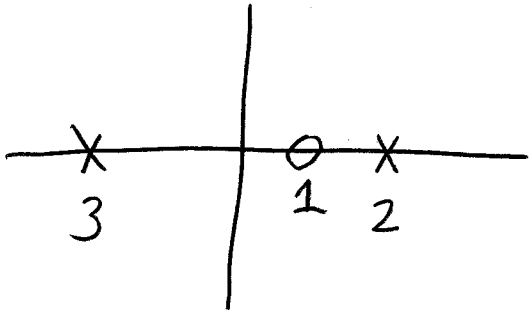


EX:

(6-15)

$$\text{Let } X(z) = \frac{z-1}{(z-2)(z+3)}$$

P-Z plot



- if  $x[n]$  is right-sided, the ROC must be  $|z| > 3$ .  $X(e^{j\omega})$  does not exist in this case, since the ROC does not include the unit circle.

- if  $x[n]$  is left-sided, the ROC must be  $|z| < 2$ .  $X(e^{j\omega})$  does exist in this case, since the ROC does include the unit circle.

- if  $x[n]$  is two-sided, the ROC must be  $2 < |z| < 3$ .  $X(e^{j\omega})$  does not exist in this case, since the ROC does not include the unit circle.

(6-16)  
- In some applications, you may be interested only in causal signals  $x[n]$  for which  $x[n] = 0 \quad \forall n < 0$  and causal systems  $H$  for which  $h[n] = 0 \quad \forall n < 0$ .

- In such cases, you can use the unilateral z-transform

$$X_u(z) = \sum_{n=0}^{\infty} x[n] z^{-n}$$

- With the unilateral z-transform, you don't have to specify the ROC.

→ It's always the exterior of the circle that passes through the largest pole.

NOTE: the z-transform that is implemented in Matlab is the unilateral z-transform.

NOTE: the more general z-transform

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$

is called the bilateral z-transform when there is a need to distinguish it from the unilateral z-transform.

(6-17)

- When you are asked to compute the z-transform  $X(z) = \mathcal{Z}\{x[n]\}$ ,

→ Your first hope is to find  $x[n]$  on a table of z-transforms like the one on the course web site or the one in Table 6.1 of the text.

→ If  $x[n]$  is not in your table, then your second hope is that you can use z-transform properties to make the one you've got look like one that's in the table.

→ Tables of z-transform properties can be found on the course web site and in Table 6.2 of the text.

→ If both of the above fail, then you will have to try to calculate  $X(z)$  from the definition.

NOTE: if  $x[n] \in \mathcal{L}'(\mathbb{Z})$ , then  $X(e^{j\omega})$  exists.

- This means that the ROC of  $X(z)$  includes the unit circle,

- This means that all of the transform pairs in your DTFT table can be converted into z-transform pairs by making the substitution  $z = e^{j\omega}$ .

# Inversion

(6-18)

- Theoretically,  $x[n]$  can be recovered from  $X(z)$  by  $x[n]r^{-n} = \text{DTFT}^{-1}\{X(re^{j\omega})\}$ , which requires integrating around the circle of radius  $r$  (from  $\omega = -\pi$  to  $\pi$ ) for any  $r$  that is in the ROC of  $X(z)$ .

- But this involves solving a complex line integral in the complex plane, which can be involved.

- Your first hope is to find your  $X(z)$  in a table of z-transforms... then you just write down the answer.

- If the  $X(z)$  you've got isn't in the table, then try to use z-transform properties to make the one you've got look like a combination of ones that are in the table.

- If both of the above fail, you can try to manipulate  $X(z)$  into the form  $X(z) = \sum_{n=-\infty}^{\infty} x[n]z^{-n}$ . Then you can just "pick off" the values of  $x[n]$ .

EX:  $X(z) = z^2(1 - \frac{1}{2}z^{-1})(1+z^{-1})(1-z^{-1})$

(6-19)

all  $z$  except  $z=0$  and  $z=\infty$ .

$$X(z) = (1 - \frac{1}{2}z^{-1})(z+1)(z-1)$$

$$= (1 - \frac{1}{2}z^{-1})(z^2 - 1)$$

$$= z^2 - \frac{1}{2}z - 1 + \frac{1}{2}z^{-1}$$

$$= \underbrace{1z^2}_{X[-2]} - \underbrace{\frac{1}{2}z^1}_{X[1]} - \underbrace{1z^0}_{X[0]} + \underbrace{\frac{1}{2}z^{-1}}_{X[1]}$$

$$\Rightarrow X[n] = \delta[n+2] - \frac{1}{2}\delta[n+1] - \delta[n] + \frac{1}{2}\delta[n-1].$$

EX:  $X(z) = \log(1+az^{-1}), |z| > |a|.$

Expanding in a Taylor series,

$$X(z) = \sum_{n=1}^{\infty} \frac{(-1)^{n+1} a^n}{n} z^{-n}$$

$$X[n] = \begin{cases} (-1)^{n+1} \frac{a^n}{n}, & n \geq 1 \\ 0, & n < 1. \end{cases}$$

- If all of the above fail and  $X(z)$  (6-20) is rational in  $z^{-1}$  (equivalent: rational in  $z$ ), then you can try to invert by long division.

→ see the text and/or the 3793 lecture notes for examples.

- If all of the above fail, you will have to try to invert from the definition by performing complex line integration,

→ since there is no fundamental theorem of calculus for complex valued functions of a complex variable, this will generally involve an application of the residue theorem and/or one or more Cauchy integral theorems.

- Suppose that  $H$  is an LTI system that is both stable and causal.

6-21

→ Since  $H$  is BIBO stable,  $h[n] \in \ell^1(\mathbb{Z})$ .

→ This means  $H(e^{j\omega})$  exists.

→ This means the unit circle is in the ROC of  $H(z)$ .

→ Since  $H$  is causal,  $h[n] = 0 \quad \forall n < 0$ .

→ This means  $h[n]$  is right sided or finite support.

→ This means the ROC of  $X(z)$  is exterior or else is the whole  $z$ -plane (e.g., the exterior of the circle of radius zero).

→ So the ROC of  $H(z)$  includes the unit circle and is exterior.

⇒ This means that all the poles have to be inside the unit circle for a causal, stable LTI system.